

Introduction to Medical Imaging

Lecture 4: Sampling

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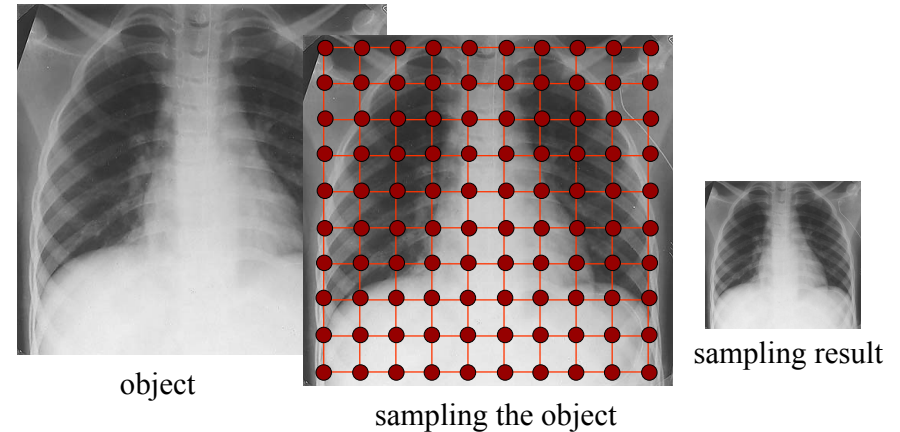
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Introduction

Sampling is the process of discretizing a continuous function into an array/matrix of data points

- the matrix values are some function of the sampled real-life object
- this function is given by the *sampling filter* (more to follow)



Importance of the Fourier Domain

Visual artifacts are also often easier understood in the Fourier domain

We can use the Fourier domain to:

- gain insight into the spatial / temporal frequency content of the data (see last lecture)
- from this, gain insight into how much a continuous signal must be sampled when it is discretized
- design proper filters to avoid an important phenomenon: *aliasing*

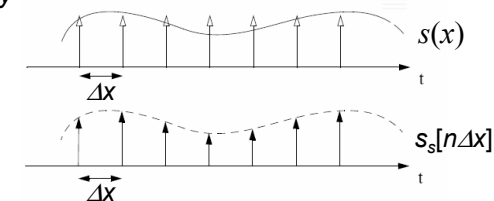
We usually do not use the Fourier domain to:

- perform the actual signal filtering, sampling, resampling, reconstruction (there are exceptions, however)
- these real operations are usually performed in the original signal domain (spatial, temporal)

Sampling: Spatial Domain

Definition:

- a continuous signal $s(x)$ is measured at fixed instances spaced apart by an interval Δx
- the data points so obtained form a discrete signal $s_s[n\Delta x] = s_s(n\Delta x)$
- here, Δx is called the *sampling period (distance)*, and $K = 1/\Delta x$ the *sampling frequency*



Sampling is the multiplication of the signal with an impulse train:

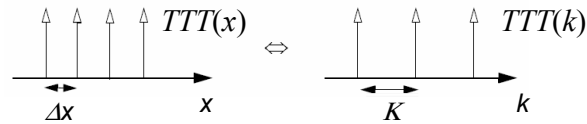
$$s_s(x) = s(x) \cdot \text{TTT}(x)$$

$$\text{TTT}(x) = \sum_{n=-\infty}^{+\infty} \delta(x - n\Delta x), \quad \text{TTT}(x) \text{ is the comb function}$$

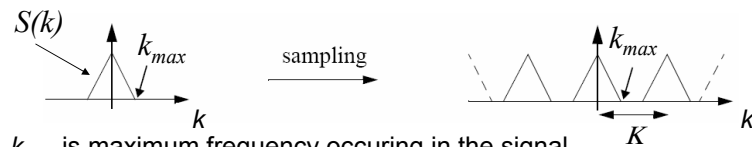
Sampling: Frequency Domain

Using the convolution theorem of the Fourier transform:

$$S_s(k) = S(k) * F\{TTT(x)\}, \text{ where } F\{TTT(x)\} = K \sum_{l=-\infty}^{+\infty} \delta(k - lK)$$



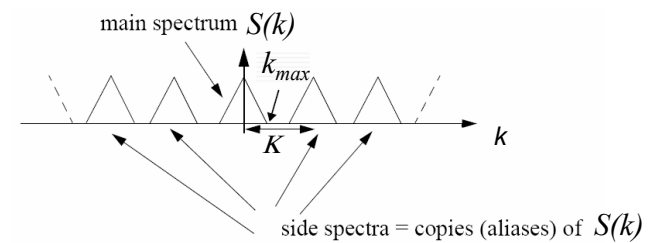
- the smaller Δx the wider K (recall the Fourier scaling theorem)
- sampling (the convolution of $TTT(k)$ and $S(k)$) replicates the signal spectrum $S(k)$ at integer multiples of sampling frequency K



- k_{max} is maximum frequency occurring in the signal

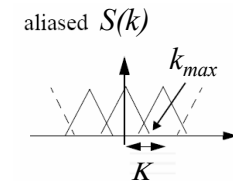
Aliasing

Terminology:



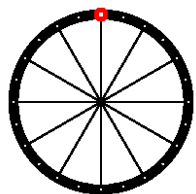
However, if we choose $K < 2 k_{max}$ the aliases overlap and we get **aliasing**

- what does aliasing look like?
- let's see some examples



Aliasing: A Commonly Observed Phenomenon

Ever wondered about the wagon wheels in old Western movies:

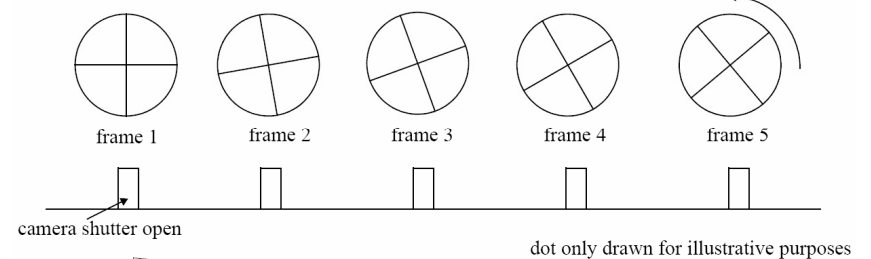


Aliasing: A Commonly Observed Phenomenon

- Wagon wheel in old Western movies:



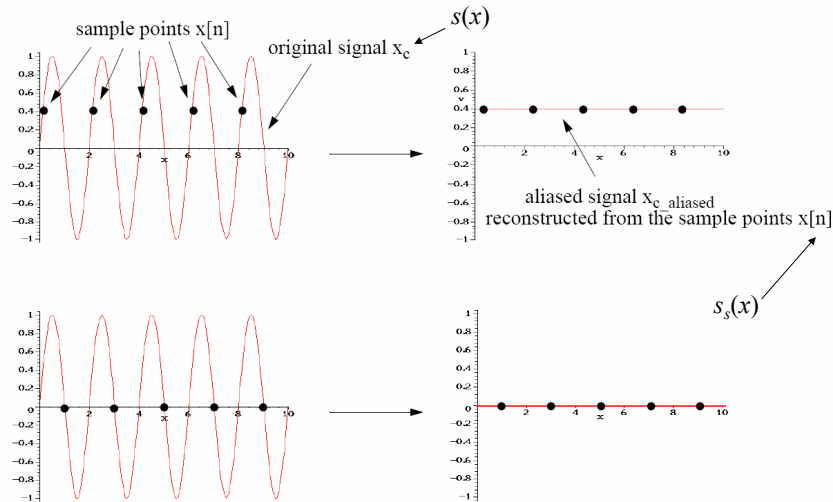
wheel appears to turn counter-clockwise at a slow rate...



but in reality turns clockwise at a much faster rate...

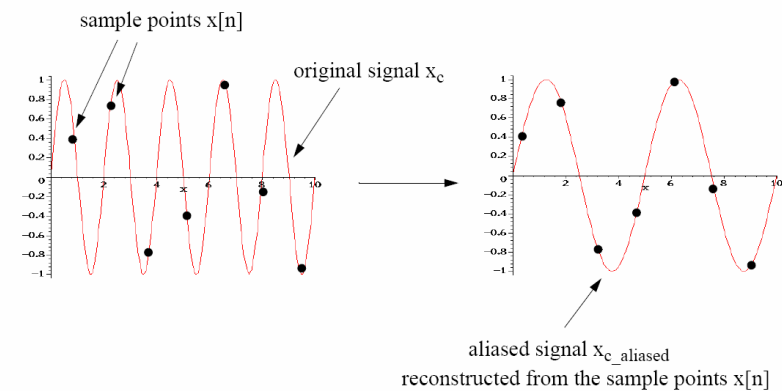
Aliasing: A More Analytical Example (1)

- Frequency of original signal: 0.5 (oscillations per time unit)
- Sampling frequency: 0.5 (samples per time unit) → original signal can not be recovered



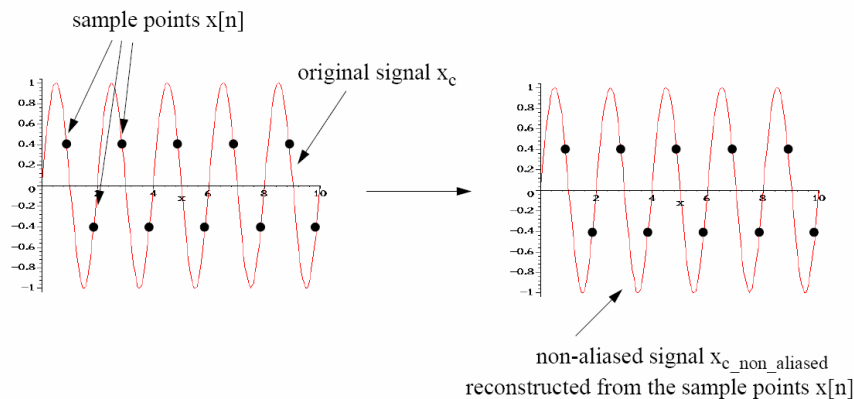
Aliasing: A More Analytical Example (2)

- Frequency of original signal: 0.5 (oscillations per time unit)
- Sampling frequency: 0.7 (samples per time unit)
- Looking at the sample points $x[n]$, they appear to originate from a sine wave $x_{c_aliased}$ of much lower frequency → again, the original sine wave is lost and can not be recovered



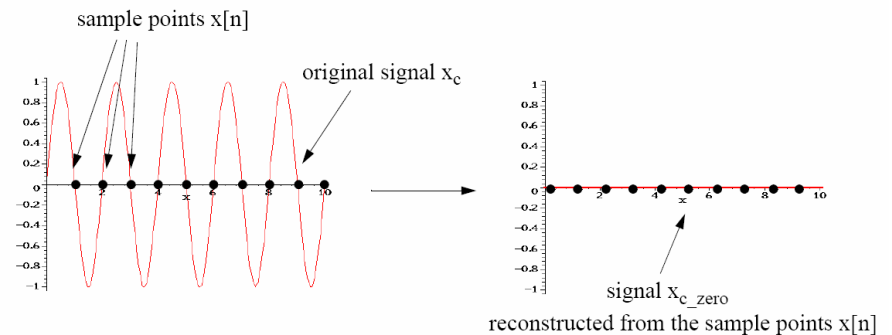
Aliasing: A More Analytical Example (3)

- Frequency of original signal: 0.5 (oscillations per time unit)
- Sampling frequency: 1.0 (sample per time unit) → original signal can be recovered
- We learn that we need to sample each oscillation period twice for good reconstruction



Aliasing: A More Analytical Example (4)

- In practice, it is best to use more than 2 samples per oscillation period
 - else one may get wrong reconstructions for some special sample alignments



- Thus, to be on the safe side:
 - sample each oscillation period more than twice
- Next: a closer look onto the whole process

Aliasing: Prevention

So must choose:

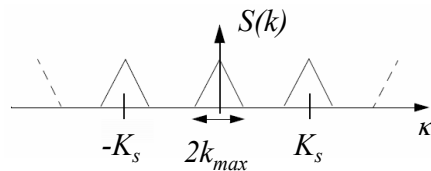
$$K > K_s = 2 \cdot k_{\max}, \quad K_s \text{ is the Nyquist rate}$$

In other words:

- the samples only uniquely define the signal if:

$$S(k) = 0 \quad \forall |k| > k_{\max}$$

$$\frac{1}{\Delta x} > 2k_{\max} = K_s$$



- this assumes that the signal is band-limited ($S(k)=0$ above K_s)

Anti-Aliasing

Usually signals are not band-limited

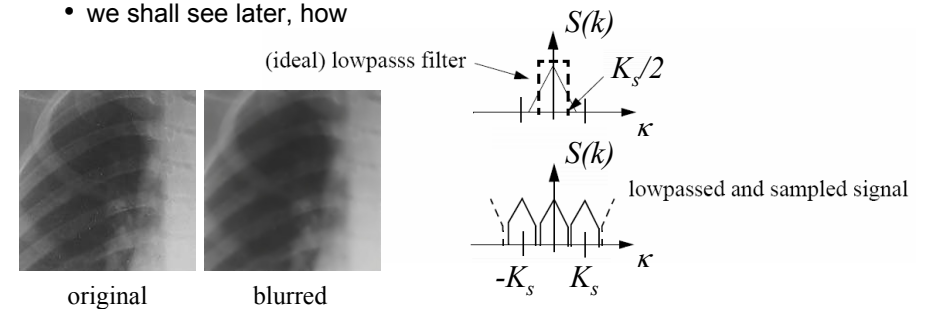
- recall the infinite spectrum of a sharp edge (for example: a bone)

To prevent the inevitable aliasing we must perform anti-aliasing before sampling the signal

- for example: when digitizing a radiograph of a bone or a chest

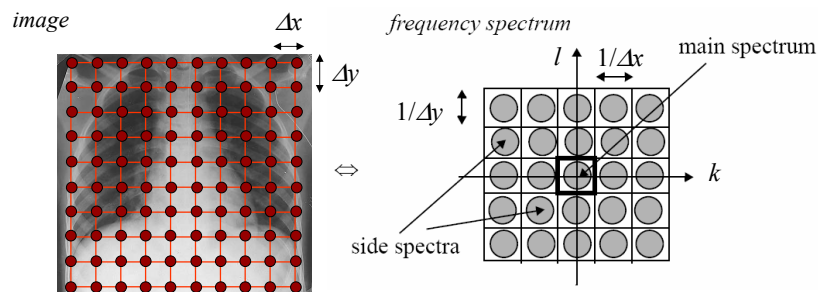
Anti-aliasing is done by low-pass filtering (blurring)

- band-limit the signal *prior* to sampling
- we shall see later, how



Higher Dimensions

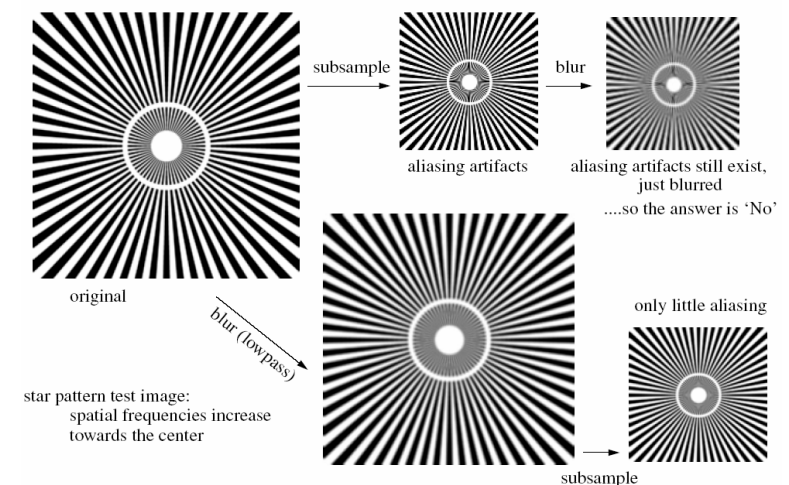
All of these concepts readily extend to higher dimensions



Main spectrum ($S(k,l)$) must fit into the center box to prevent overlap with side-spectra (and aliasing)

$$\frac{1}{\Delta x} > 2 \cdot k_{x \max} \quad \frac{1}{\Delta y} > 2 \cdot k_{y \max}$$

Anti-Aliasing: Practical Examples (1)



Anti-Aliasing: Practical Examples (2)

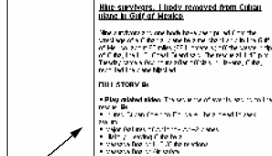
Nine survivors, 1 body removed from Cuban plane in Gulf of Mexico

Nine survivors and one body have been pulled from the wreckage of a Cuban airplane by a merchant ship in the Gulf of Mexico, about 60 miles (96 kilometers) off the western tip of Cuba, the U.S. Coast Guard said. The rescue at 1:45 p.m. Tuesday came a few hours after officials in Havana, Cuba, reported the plane hijacked.

FULL STORY

- Play related video: The sequence of events leading to the rescue
- Injured Cuban flown to Florida will be allowed to seek asylum
- Major features of Antonov An-2 planes
- History: Leaving Cuba by air
- Message Board: U.S./Cuba relations
- Message Board: Air safety

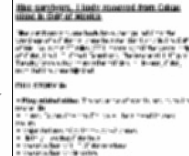
original



subsample

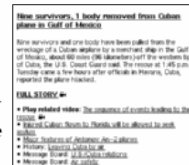
aliased text

blur



blurred, aliased text

blur, then subsample



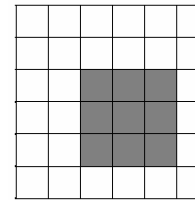
looks more pleasing

We observe: Anti-aliasing (i.e., blurring, lowpassing) must be applied before sampling

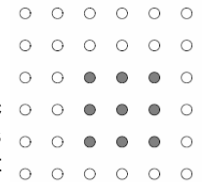
Image Representation

We know that a discrete image is a matrix of pixels

- do keep this in mind, however:



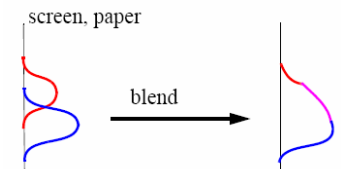
an image is NOT a matrix of solid squares



rather, each pixel is a Dirac impulse, with the pixel's value as its height

So, why do we not see isolated dots on the screen or paper?

- a monitor or printer "splats" the pixels onto the screen or paper.
- each pixels assumes the shape of a Gaussian



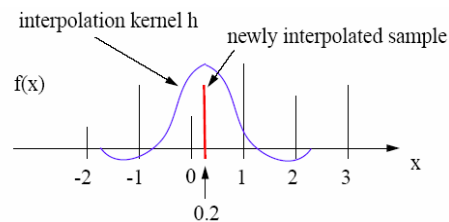
- the Gaussians blend together and form a continuous image

Interpolation

Often we want to estimate the formerly continuous function from the discretized function represented by the matrix of sample points

This is done via *interpolation*

Concept:

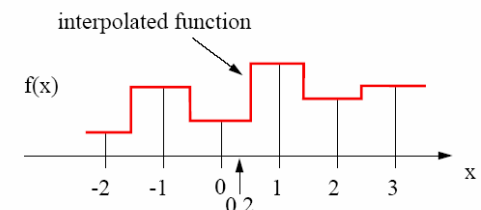
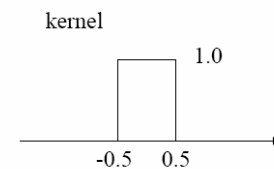


- center the interpolation kernel (filter) h at the sample position and superimpose it onto the grid
- multiply the values of the grid samples with the kernel value at the superimposed position
- add all the products $\rightarrow$ this gives the value of the newly interpolated sample
- in the shown case:

$$f(0.2) = h(-0.2) f(0) + h(-1.2) f(-1) + h(0.8) f(1) + h(1.8) f(2)$$

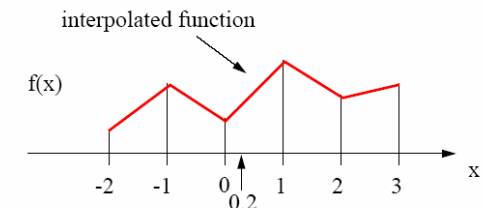
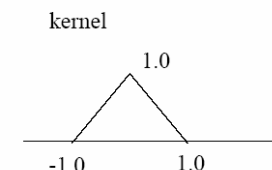
Interpolation Kernels (1)

- Nearest Neighbor:



- simply pick the value of the nearest grid point: $f(0.2) = f(\text{trunc}(0.2+0.5)) = f(\text{round}(0.2))$

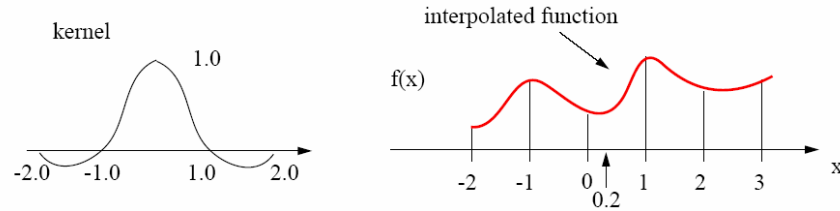
- Linear filter:



- use a linear combination of the two neighboring grid values: $f(0.2) = 0.2 \cdot f(1) + 0.8 \cdot f(0)$

Interpolation Kernels (2)

- Cubic filter:



An additional popular filter is the Gaussian function

Discussion:

- nearest neighbor is fastest to compute (just one add), gives sharp edges, but sometimes jagged lines
- linear interpolation takes 2 mults and 1 add and gives a piecewise smooth function
- cubic filter takes 4 mults and 3 adds, but gives an overall smooth interpolated function
- linear interpolation is most popular in many application

Interpolation in Higher Dimensions

- All interpolation kernels shown here are separable

$$h(x, y) = h(x) \cdot h(y) \quad \text{and} \quad h(x, y, z) = h(x) \cdot h(y) \cdot h(z)$$

- Linear interpolation

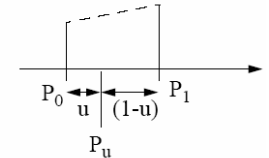
assume: grid distance = 1.0

P_u is the location of the sample value

P_0 and P_1 are neighboring grid points

then: $u = P_u - P_0$

$$f(x) = f(P_u) = (1 - u) \cdot f(P_0) + u \cdot f(P_1)$$



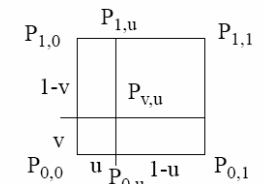
- Bilinear interpolation

$$f(P_{0,u}) = (1 - u) \cdot f(P_{0,0}) + u \cdot f(P_{0,1})$$

$$f(P_{1,u}) = (1 - u) \cdot f(P_{1,0}) + u \cdot f(P_{1,1})$$

$$f(P_{v,u}) = (1 - v) \cdot f(P_{0,u}) + v \cdot f(P_{1,u})$$

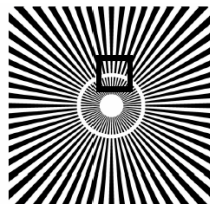
$$\rightarrow f(x, y) = f(P_{v,u}) = (1-v)(1-u)f(P_{0,0}) + (1-v)uf(P_{0,1}) + v(1-u)f(P_{1,0}) + vuf(P_{1,1})$$



Interpolation Quality

Example:

- resampling of a portion of the star image onto a high resolution grid
- magnification factor ~20



Computation of the Fourier Transform

The analytical form of the Fourier transform (and its laws) is convenient for theoretical, fundamental considerations

- examples: filter design, sampling rates, image resolutions

But in practical applications (for example, low-passing and other filtering) we require a means to compute a discretized signal's Fourier transform:

$$S(m\Delta k_x, n\Delta k_y) = \sum_{q=0}^{N-1} \sum_{p=0}^{M-1} s(p\Delta x, p\Delta y) e^{-2\pi i \left(\frac{mp}{M} + \frac{nq}{N} \right)}$$

$$s(p\Delta x, q\Delta y) = \sum_{n=0}^{N-1} \sum_{m=0}^{M-1} S(m\Delta k_x, n\Delta k_y) e^{2\pi i \left(\frac{mp}{M} + \frac{nq}{N} \right)}$$

Assume $M=N$, then this is an $O(N^4)$ algorithm

- the Fast Fourier Transform (FFT) brings this down to $O(N^2 \log N)$